

Indian Institute of Science  
 E9–252: Mathematical Methods and Techniques in Signal Processing  
 Instructor: Shayan G. Srinivasa  
 Homework #3 Solutions, Fall 2017

Solutions prepared by Priya J Nadkarni

Late submission policy: Points scored = Correct points scored  $\times e^{-d}$ ,  $d = \#$  days late

**Assigned date:** Sept. 11<sup>th</sup> 2017      **Due date:** Sept. 18<sup>th</sup> 2017 by end of the day

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**PROBLEM 1:**

Solve problems 4.6 and 4.11 from P. P. Vaidyanathan's book. (8 + 10 points)

**Solution:**

(4.6) For the first system, we obtain,

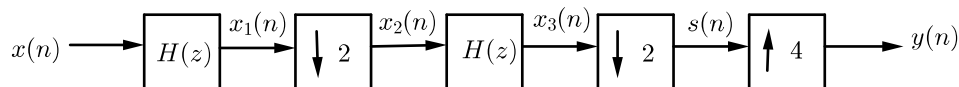
$$\begin{aligned}
 y_0(n) &= h_k(n) * y(n) \Rightarrow Y_0(z) = H_k(z)Y(z) \\
 h_k(n) &= h_0(n) \cos \frac{2\pi kn}{L} = \frac{h_0(n)e^{j\frac{2\pi kn}{L}} + h_0(n)e^{-j\frac{2\pi kn}{L}}}{2} \\
 H_k(z) &= \frac{H_0(e^{-j\frac{2\pi k}{L}}z) + H_0(e^{j\frac{2\pi k}{L}}z)}{2} \\
 Y_0(z) &= \frac{H_0(e^{-j\frac{2\pi k}{L}}z)Y(z) + H_0(e^{j\frac{2\pi k}{L}}z)Y(z)}{2}
 \end{aligned}$$

For the second system,  $y_1(n) = (h_0(n) * y(n)) \cos \frac{2\pi kn}{L}$ . Let  $x(n) = h_0(n) * y(n)$ , then,

$$\begin{aligned}
 X(z) &= H_0(z)Y(z) \\
 Y_1(z) &= \frac{X(e^{-j\frac{2\pi k}{L}}z) + X(e^{j\frac{2\pi k}{L}}z)}{2} \\
 &= \frac{H_0(e^{-j\frac{2\pi k}{L}}z)Y(e^{-j\frac{2\pi k}{L}}z) + H_0(e^{j\frac{2\pi k}{L}}z)Y(e^{j\frac{2\pi k}{L}}z)}{2}
 \end{aligned}$$

$Y_0(z) = Y_1(z)$  only if  $Y(e^{-j\frac{2\pi k}{L}}z) = Y(z)$ . Otherwise,  $Y_1(z) \neq Y_0(z)$ . One such case where they are equal is when  $k \bmod L = 0$ . Another example where it is equal is when the signal has been upsampled by L (Refer Prob 4.5). Thus, they are not necessarily the same. They may be same in some cases, but differ in others.

(4.11) Consider the following system:



When we pass  $x(n)$  through the filter  $H(z)$ , we obtain

$$X_1(z) = X(z)H(z)$$

When we downsample  $x_1(n)$  by 2, we obtain,

$$X_2(z) = \frac{X_1(z) + X_1(-z)}{2}$$

When we pass  $x_2(n)$  through the filter  $H(z)$ , we obtain,

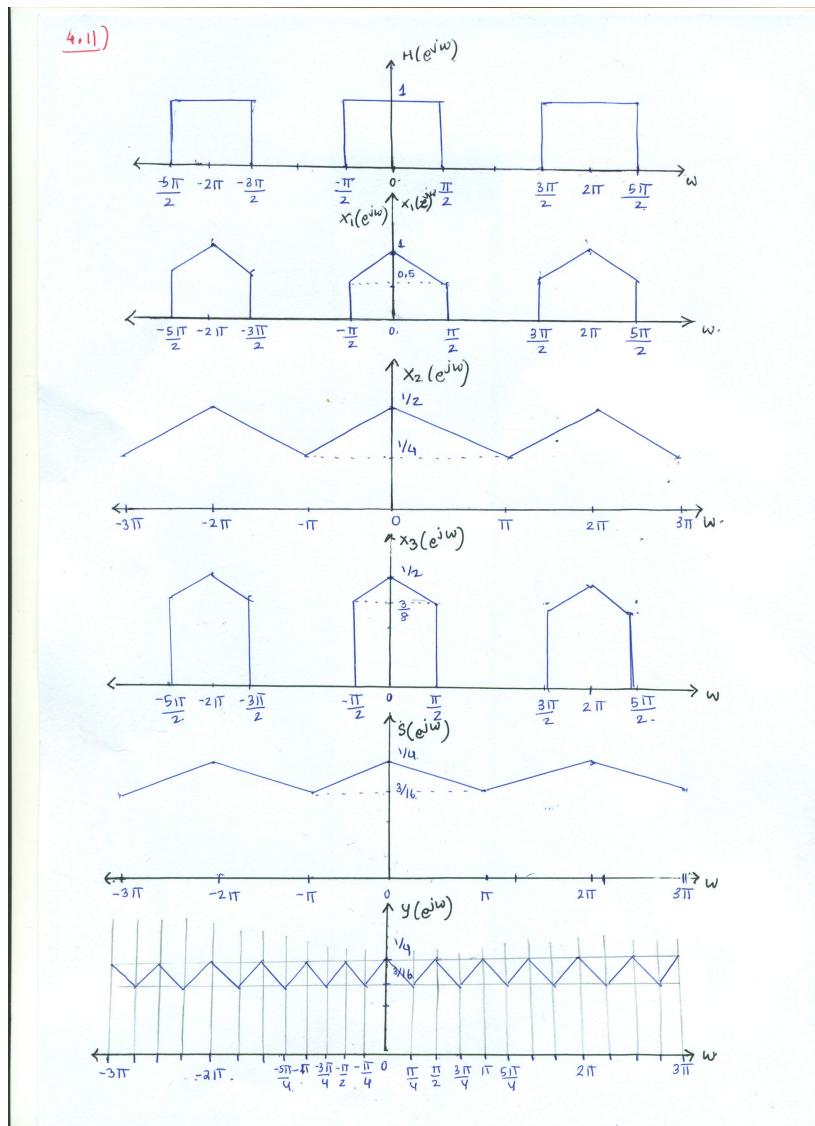
$$X_3(z) = X_2(z)H(z)$$

When we downsample  $x_3(n)$  by 2, we obtain,

$$S(z) = \frac{X_3(z) + X_3(-z)}{2}$$

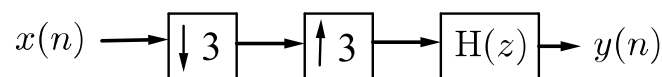
When we upsample  $s(n)$  by 4, we obtain,

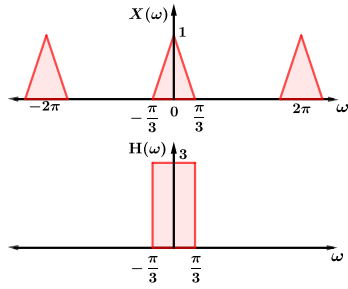
$$Y(z) = S(z^4)$$



**PROBLEM 2:**

Consider the following system:

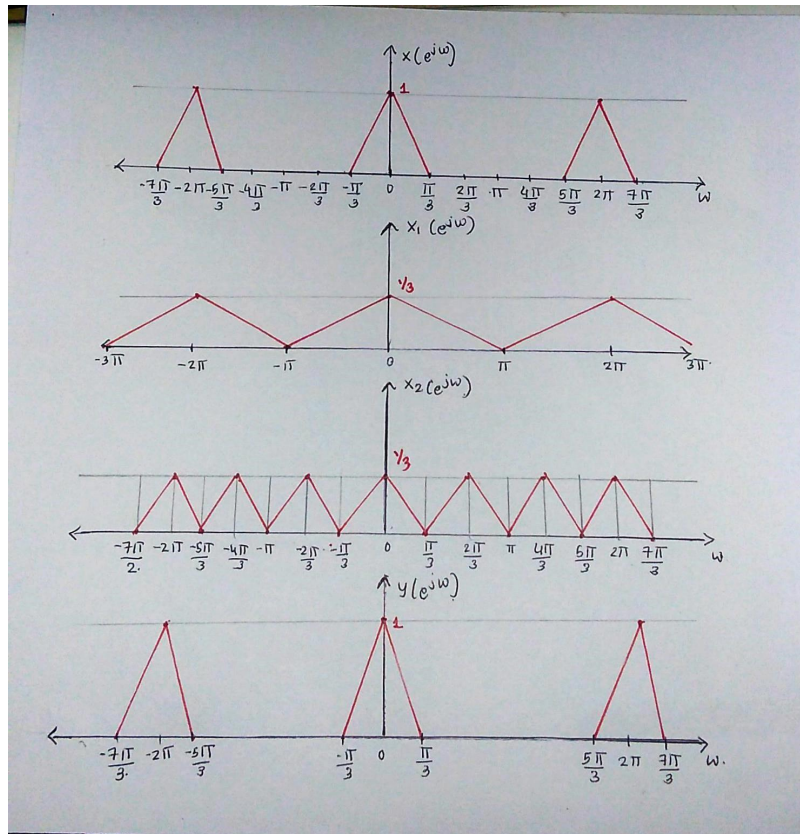




Suppose the spectrum of the original signal and transfer function is as given above.

Analyze the spectrum of  $y(t)$ . Analyze the output spectrum if the decimator and expander and interchanged. (4+3 points)

**Solution:** Let  $x_1(n)$  and  $x_2(n)$  denote the intermediate signals.



If the decimator and the interpolator are exchanged,

$$\begin{aligned}
 X_1(z) &= X(z^3) \\
 X_2(z) &= \frac{1}{3} \sum_{k=0}^2 X_1(z^{1/3} e^{-j\frac{2\pi k}{3}}) \\
 &= \frac{1}{3} \sum_{k=0}^2 X((z^{1/3} e^{-j\frac{2\pi k}{3}})^3) = \frac{1}{3} \sum_{k=0}^2 X(z) = X(z) \\
 Y(z) &= 3X(z)
 \end{aligned}$$

The last equality follows as  $X(z)$  is bandlimited to  $\frac{\pi}{3}$ .